#### **Frequency Response of LTI Systems**

Digital Signal Processing

October 21, 2025



#### **Review: Transfer Function**

Given an LTI system with impulse response h[n]:

$$y[n] = x[n] * h[n].$$

The **transfer function** is the z-transform of h[n]. It is given by the ratio:

$$H(z) = \frac{Y(z)}{X(z)},$$

where  $x[n] \overset{\mathcal{Z}}{\longleftrightarrow} X(z)$  and  $y[n] \overset{\mathcal{Z}}{\longleftrightarrow} Y(z)$ .

#### **Review: Transfer Function of LCCDE**

A linear constant-coefficient difference equation (LCCDE) is an LTI system of the form:

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k].$$

It's transfer function is

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^{M} b_k z^{-k}}{\sum_{k=0}^{N} a_k z^{-k}}.$$

#### **Exercise: Moving Average**

What is the transfer function for the following moving average system?

$$y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k].$$

### **Answer: Moving Average**

Moving Average:

$$y[n] = \frac{1}{M} \sum_{k=0}^{M-1} x[n-k].$$

Taking *z*-transform of both sides:

$$Y(z) = \frac{1}{M} \sum_{k=0}^{M-1} z^{-k} X(z).$$

Solving for the transfer function:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{1}{M} \sum_{k=0}^{M-1} z^{-k}.$$

#### **Pole-Zero Plot for Moving Average**

Rearrange the transfer function:

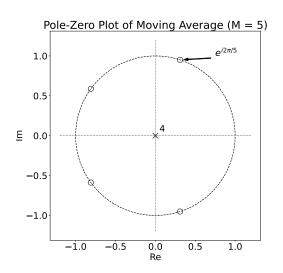
$$H(z) = \frac{1}{M} \sum_{k=0}^{M-1} z^{-k}$$
$$= \frac{1}{M} \frac{\sum_{k=0}^{M-1} z^k}{z^{M-1}}$$

multiply by  $\frac{z^{M-1}}{z^{M-1}}$ 

Trick: 
$$(1 + z + z^2 + \dots + z^{M-1})(z - 1) = z^M - 1$$
.

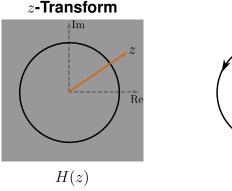
So,  $\sum_{k=0}^{M-1} z^k = \frac{z^M-1}{(z-1)}$ , which means the zeros are the Mth roots of 1, excluding z=1.

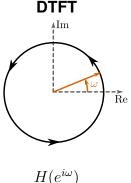
### Pole-Zero Plot for Moving Average



#### Frequency Response of an LTI System

The **frequency response** of an LTI system is the restriction of H(z) to the unit circle, which is the DTFT of the impulse response,  $H(e^{i\omega})$ .





$$\begin{split} T\{e^{i\omega n}\} &= e^{i\omega n} * h[n] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega(n-k)} h[k] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega n} e^{-i\omega k} h[k] \\ &= e^{i\omega n} \sum_{k=-\infty}^{\infty} e^{-i\omega k} h[k] & \text{DTFT of } h[k]! \\ &= H(e^{i\omega}) e^{i\omega n} \end{split}$$

$$\begin{split} T\{e^{i\omega n}\} &= e^{i\omega n} * h[n] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega(n-k)} h[k] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega n} e^{-i\omega k} h[k] \\ &= e^{i\omega n} \sum_{k=-\infty}^{\infty} e^{-i\omega k} h[k] & \text{DTFT of } h[k]! \\ &= H(e^{i\omega}) e^{i\omega n} \end{split}$$

$$\begin{split} T\{e^{i\omega n}\} &= e^{i\omega n} * h[n] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega(n-k)} h[k] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega n} e^{-i\omega k} h[k] \\ &= e^{i\omega n} \sum_{k=-\infty}^{\infty} e^{-i\omega k} h[k] & \text{DTFT of } h[k]! \\ &= H(e^{i\omega}) e^{i\omega n} \end{split}$$

$$\begin{split} T\{e^{i\omega n}\} &= e^{i\omega n} * h[n] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega(n-k)} h[k] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega n} e^{-i\omega k} h[k] \\ &= e^{i\omega n} \sum_{k=-\infty}^{\infty} e^{-i\omega k} h[k] & \text{DTFT of } h[k]! \\ &= H(e^{i\omega}) e^{i\omega n} \end{split}$$

$$\begin{split} T\{e^{i\omega n}\} &= e^{i\omega n} * h[n] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega(n-k)} h[k] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega n} e^{-i\omega k} h[k] \\ &= e^{i\omega n} \sum_{k=-\infty}^{\infty} e^{-i\omega k} h[k] & \text{DTFT of } h[k]! \\ &= H(e^{i\omega}) e^{i\omega n} \end{split}$$

Consider a complex sinusoid  $e^{i\omega n}$ . Applying  $T\{\cdot\}$  to it gives:

$$\begin{split} T\{e^{i\omega n}\} &= e^{i\omega n} * h[n] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega(n-k)} h[k] \\ &= \sum_{k=-\infty}^{\infty} e^{i\omega n} e^{-i\omega k} h[k] \\ &= e^{i\omega n} \sum_{k=-\infty}^{\infty} e^{-i\omega k} h[k] \end{split} \qquad \text{DTFT of } e^{i\omega n} \\ &= H(e^{i\omega}) e^{i\omega n} \end{split}$$

DTFT of h[k]!

#### **Frequency Response**

Let  $T\{\cdot\}$  be an LTI system. Recall it's **impulse response** is

$$h[n] = T\{\delta[n]\}.$$

It's **frequency response** is how it responds to a complex sinusoid with a certain frequency  $\omega \in [0, 2\pi)$ :

$$T\{e^{i\omega n}\} = H(e^{i\omega})e^{i\omega n}.$$

## Magnitude and Phase of Frequency Response

Looking at frequency response of an LTI:

$$Y(e^{i\omega}) = H(e^{i\omega})X(e^{i\omega})$$

Remember complex multiplication in Euler form:

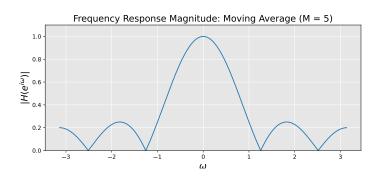
$$re^{i\theta} \cdot se^{i\phi} = (rs)e^{i(\theta+\phi)}$$

So, we have:

Magnitude:  $|Y(e^{i\omega})| = |H(e^{i\omega})| \cdot |X(e^{i\omega})|$ 

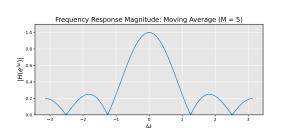
Phase:  $\operatorname{Arg}(Y(e^{i\omega})) = \operatorname{Arg}(H(e^{i\omega})) + \operatorname{Arg}(X(e^{i\omega}))$ 

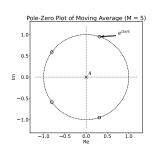
# **Frequency Response Magnitude of the Moving Average**



This is a low-pass filter.

# Frequency Response Magnitude of the Moving Average





Note the zeros.

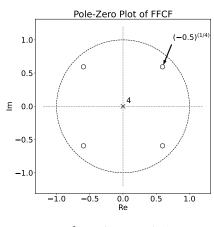
#### **Feedforward Comb Filter (FFCF)**

#### FFCF:

$$y[n] = x[n] + gx[n-k]$$

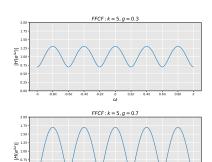
#### Transfer function:

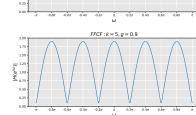
$$H(z) = \frac{z^k + g}{z^k}$$



$$k = 4, g = 0.5$$

### **FFCF Frequency Response**





FFCF: k = 5, q = 0.5

0.50

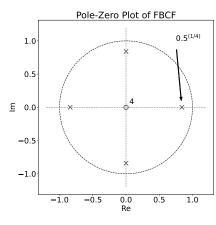
0.25

0.50

#### Feedback Comb Filter (FBCF)

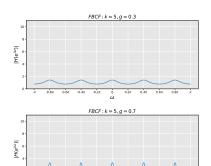
$$y[n] = x[n] + gy[n-k]$$

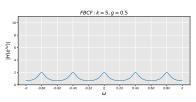
$$H(z) = \frac{z^k}{z^k - q}$$

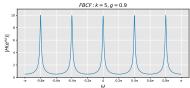


$$k = 4, g = 0.5$$

### **FBCF Frequency Response**







#### **Exponential Moving Average**

Exponential moving average filter is given by:

$$y[n] = (1 - g)x[n] + gy[n - 1],$$

where 0 < g < 1.

Note, this is the feedback comb filter with delay 1 and slightly different constants, so it has transfer function:

$$H(z) = \frac{(1-g)z}{z-g}.$$

# Frequency Response of Exponential Moving Average

